

The uniting theorem

- ❑ Key tool to simplification: $X Y' + XY = X(Y'+Y) = X(1) = X$
- ❑ Essence of simplification of two-level logic
 - Find a pair of PoS minterms where only one variable changes its value
 - Merge the two terms into one, and eliminate the changing variable

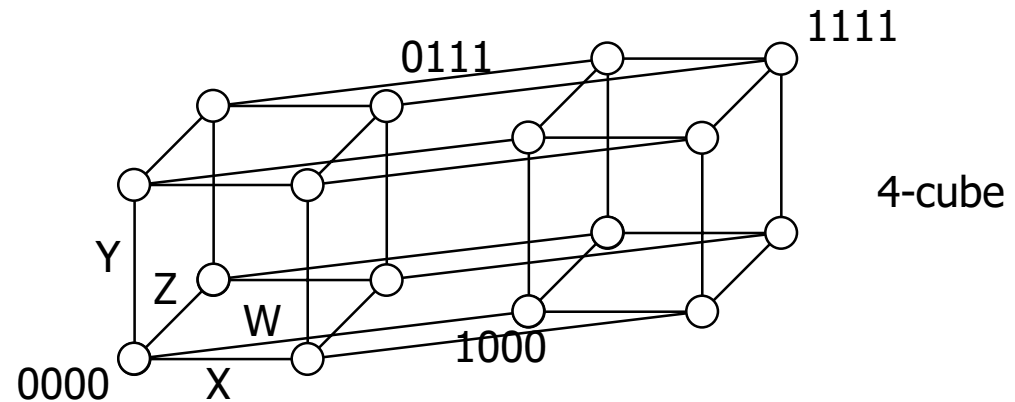
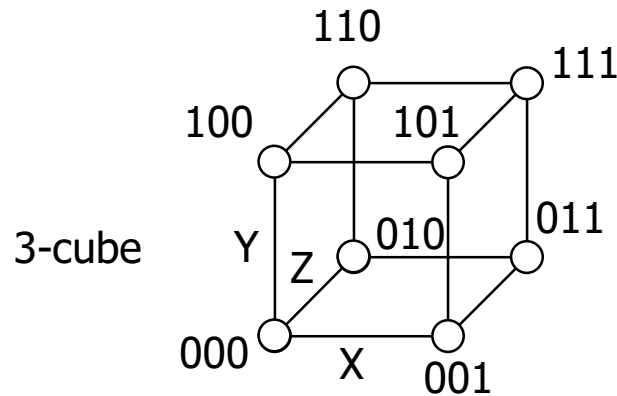
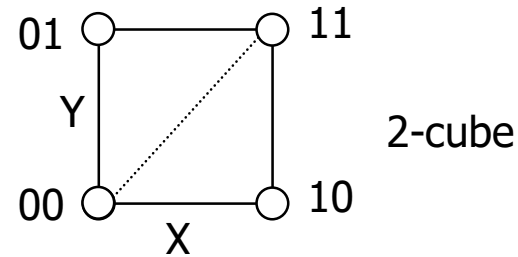
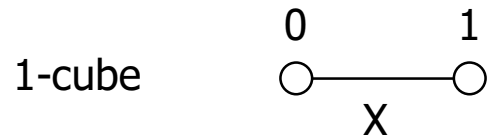
A	B	H
0	0	1
0	1	0
1	0	1
1	1	0

B has the same value in both on-set rows
– B remains (B is X)

A has a different value in the two rows
– A is eliminated (A is Y)

Boolean cubes

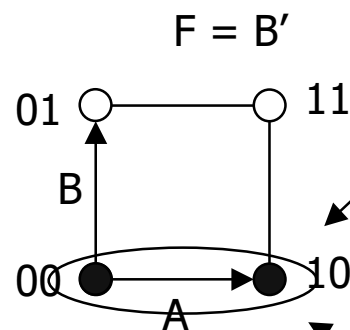
- n input variables \rightarrow one node for each minterm $\rightarrow n^2$ nodes
- Edges represent single variable changes between minterms
- A sub-cube is an m -cube, with $m < n$
- A "face" of a n -cube is an $(n-1)$ sub-cube



Mapping truth tables onto Boolean cubes

- ❑ Uniting theorem can be applied to complete faces or sub-cubes of a cube
- ❑ Example:

A	B	F
0	0	1
0	1	0
1	0	1
1	1	0



two faces of size 0 (nodes)
combine into a face of size 1 (line)

A loop enclosing a sub-cube
indicates that a single term is
needed to "cover" those minterms

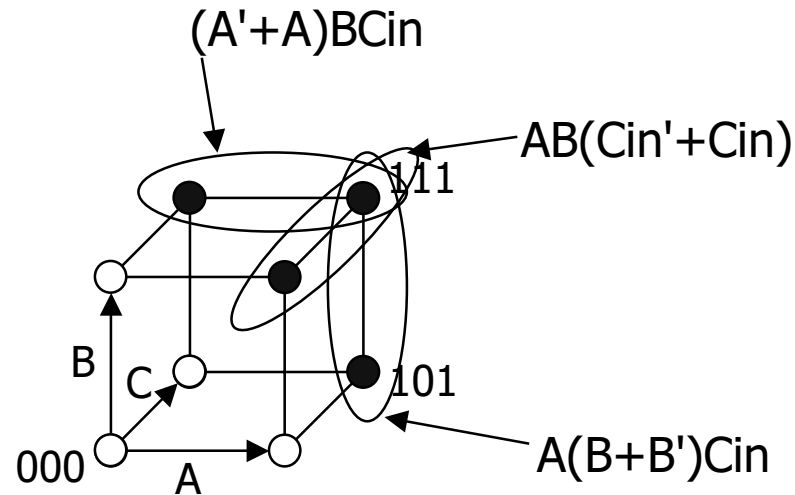
**The larger the loop, the
fewer literals in the term**

ON-set = solid nodes
OFF-set = empty nodes
DC-set = x'd nodes

Three variable example

- Binary full-adder carry-out logic

A	B	Cin	Cout
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	0
1	0	1	1
1	1	0	1
1	1	1	1



the on-set is completely covered by the combination (OR) of the sub-cubes of lower dimensionality - note that "111" is covered three times

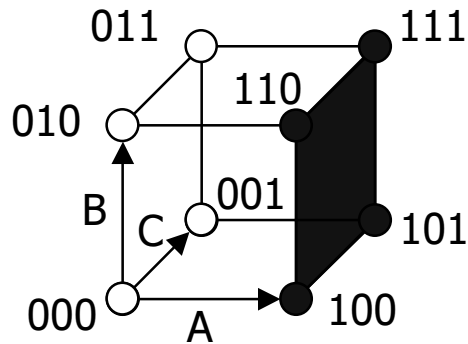
$$Cout = BCin + AB + ACin$$

Interpretation of Subcubes (Faces)

□ For a 3-cube

- A node (0-cube) is a 3-literal term (like $AB'C'$)
- An edge (1-cube) is a 2-literal term (Like AB')
- A face (2-cube) is a 1-literal term (Like A)

$$F(A,B,C) = \Sigma m(4,5,6,7)$$



A is asserted (true) and unchanged
B and C vary

□ In general

- an m -subcube within an n -cube ($m < n$) yields a term with $n - m$ literals

Karnaugh maps

- ❑ Flat map of Boolean cube
 - Cells in the table are adjacent if they vary by 1 bit
 - wrap-around at edges
 - hard to draw and visualize for more than 4 dimensions
 - virtually impossible for more than 6 dimensions
- ❑ Alternative to truth-tables to help visualize adjacencies
 - guide to applying the uniting theorem
 - on-set elements with only one variable changing value are adjacent unlike the situation in a linear truth-table

A	B	F
0	0	1
0	1	0
1	0	1
1	1	0

		A	
		0	1
B	0	1	1
	1	0	0

A 2x2 Karnaugh map for the function F = B'. The top row (B=0) contains 1s in both columns (A=0 and A=1). The bottom row (B=1) contains 0s in both columns. A horizontal oval encircles the two 1s in the top row, indicating they are adjacent and can be grouped together to form the term B'.

$$\mathbf{F = B'}$$

Karnaugh maps (cont'd)

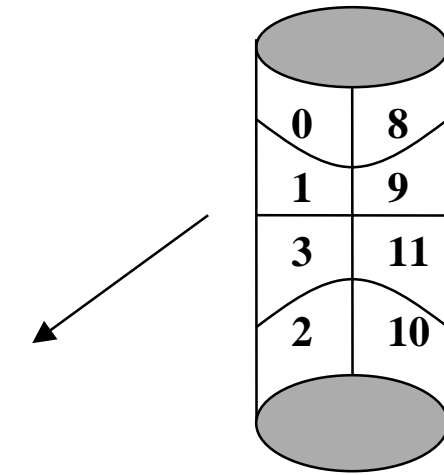
□ Numbering scheme based on Gray-code

- e.g., 00, 01, 11, 10
- only a single bit changes in code for adjacent map cells including wraparound

		AB		A	
		00	01	11	10
C	0	0	2	6	4
	1	1	3	7	5
		B			

		A			
		0	2	6	4
C	0	0	2	6	4
	1	1	3	7	5
		B			

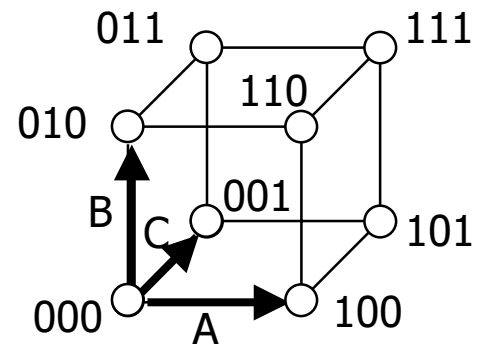
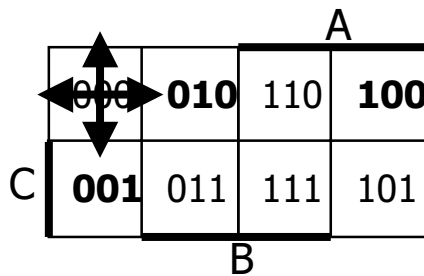
		A			
		0	4	12	8
C	0	0	4	12	8
	1	1	5	13	9
		B		D	
0	2	6	14		10
1	3	7	15	11	



13 = 1101 = ABC'D

Adjacencies in Karnaugh maps

- ❑ Wrap from first to last column
- ❑ Wrap top row to bottom row



Karnaugh map examples

□ F =

	A	
	1	1
B	0	0

B'

□ Cout =

□ f(A,B,C) = $\Sigma m(0,4,6,7)$

	A			
	0	0	1	0
Cin	0	1	1	1
	B			

AB + ACin + BCin

	A			
	1	0	0	1
C	0	0	1	1
	B			

AC + B'C' + ~~AB'~~

The consensus theorem
 AB' is "covered" by other terms

obtain the complement of the function by covering 0s with sub-cubes

More Karnaugh map examples

		A	
	0	0	1
	0	0	1
C		B	

$$G(A,B,C) = A$$

		A		
	1	0	0	1
	0	0	1	1
C		B		

$$F(A,B,C) = \sum m(0,4,5,7) = AC + B'C'$$

		A	
	1	0	0
	0	0	1
C		B	

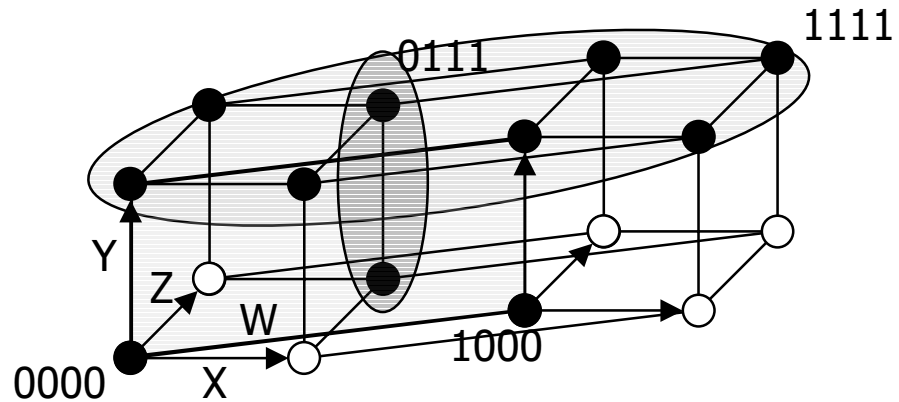
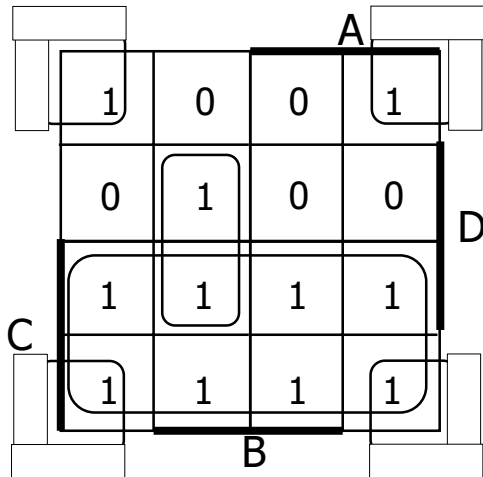
F' simply replace cover the 0's instead of the 1's

$$F'(A,B,C) = \sum m(1,2,3,6) = BC' + A'C$$

Karnaugh map: 4-variable example

□ $F(A,B,C,D) = \Sigma m(0,2,3,5,6,7,8,10,11,14,15)$

$F = C + A'BD + B'D'$



find the smallest number of the largest possible sub-cubes to cover the ON-set
(fewer terms with fewer inputs per term)

Karnaugh maps: don't cares

- $f(A,B,C,D) = \Sigma m(1,3,5,7,9) + d(6,12,13)$
 - without don't cares (don't cares become 0)
 - $f = A'D + B'C'D$

		A			
		0	0	X	0
		1	1	X	1
		1	1	0	0
		0	X	0	0
		B			
C					D

Karnaugh maps: don't cares (cont'd)

□ $f(A,B,C,D) = \Sigma m(1,3,5,7,9) + d(6,12,13)$

➤ $f = A'D + B'C'D$

without don't cares

➤ $f = A'D + C'D$

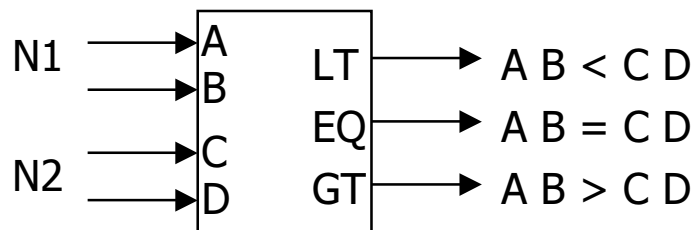
with don't cares

		A		
	0	0	X	0
	1	1	X	1
C	1	1	0	0
	0	X	0	0
		B		

This don't care completes a 2 cube

Don't cares can be treated as 1s or 0s depending on which is more advantageous

Design example: two-bit comparator

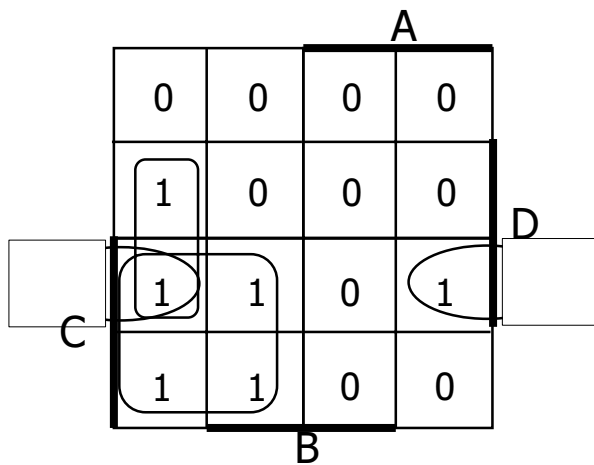


block diagram
and
truth table

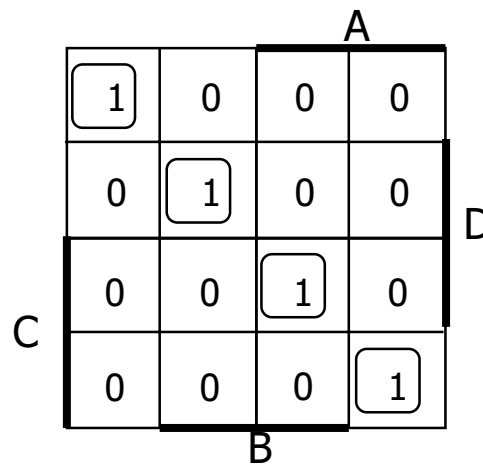
A	B	C	D	LT	EQ	GT
0	0	0	0	0	1	0
		0	1	1	0	0
		1	0	1	0	0
		1	1	1	0	0
0	1	0	0	0	0	1
		0	1	0	1	0
		1	0	1	0	0
		1	1	1	0	0
1	0	0	0	0	0	1
		0	1	0	0	1
		1	0	0	1	0
		1	1	1	0	0
1	1	0	0	0	0	1
		0	1	0	0	1
		1	0	0	0	1
		1	1	0	1	0

we'll need a 4-variable Karnaugh map
for each of the 3 output functions

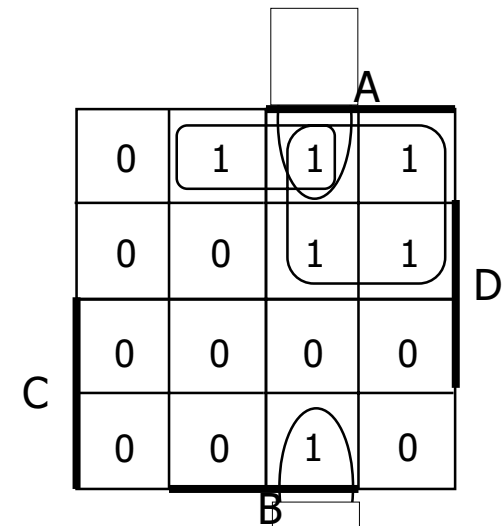
Design example: two-bit comparator (cont'd)



K-map for LT



K-map for EQ



K-map for GT

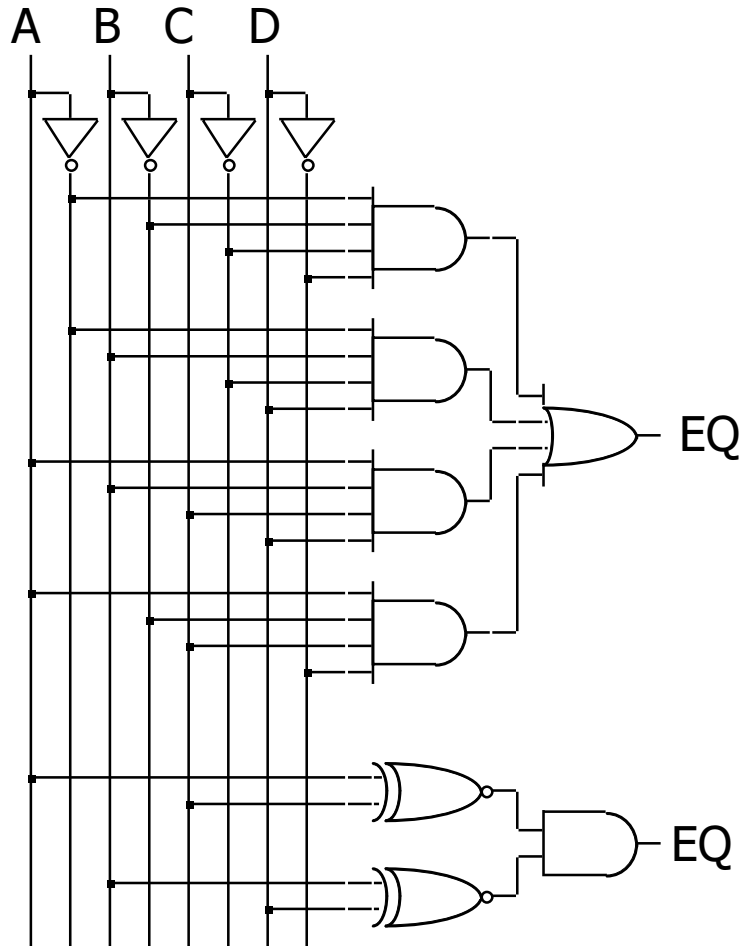
$$LT = A' B' D + A' C + B' C D$$

$$EQ = A' B' C' D' + A' B C' D + A B C D + A B' C D' = (A \text{ xnor } C) \cdot (B \text{ xnor } D)$$

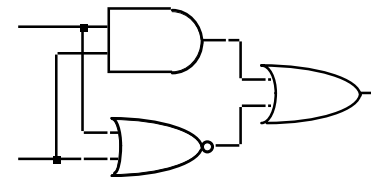
$$GT = B C' D' + A C' + A B D'$$

LT and GT are similar (flip A/C and B/D)

Design example: two-bit comparator (cont'd)

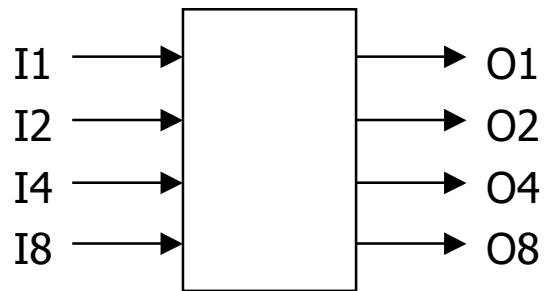


two alternative implementations of EQ with and without XNOR



XNOR is implemented with at least 3 simple gates

Design example: BCD increment by 1



block diagram
and
truth table

I8	I4	I2	I1	O8	O4	O2	O1
0	0	0	0	0	0	0	1
0	0	0	1	0	0	1	0
0	0	1	0	0	0	1	1
0	0	1	1	0	1	0	0
0	1	0	0	0	1	0	1
0	1	0	1	0	1	1	0
0	1	1	0	0	1	1	1
1	0	0	0	1	0	0	0
1	0	0	1	0	0	0	1
1	0	1	0	X	X	X	X
1	0	1	1	X	X	X	X
1	1	0	0	X	X	X	X
1	1	0	1	X	X	X	X
1	1	1	0	X	X	X	X
1	1	1	1	X	X	X	X

4-variable K-map for each of
the 4 output functions

Design example: BCD increment by 1 (cont'd)

		I8		
	0	0	X	1
	0	0	X	0
I2	0	1	X	X
	0	0	X	X
		I4		

$$O8 = I4 I2 I1 + I8 I1'$$

$$O4 = I4 I2' + I4 I1' + I4' I2 I1$$

$$O2 = I8' I2' I1 + I2 I1'$$

$$O1 = I1'$$

		I8		
	0	0	X	0
	1	1	X	0
I2	0	0	X	X
	1	1	X	X
		I4		

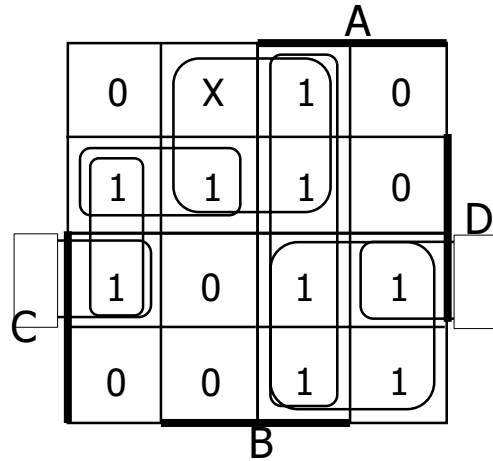
		I8		
O4	0	1	X	0
	0	1	X	0
I2	1	0	X	X
	0	1	X	X
		I4		

		I8		
O1	1	1	X	1
	0	0	X	0
I2	0	0	X	X
	1	1	X	X
		I4		

Definition of terms for two-level simplification

- ❑ Implicant
 - A subset of the on-set that combine to form a cube
- ❑ Prime implicant
 - implicant that can't be combined with another to form a larger subcube
- ❑ Essential prime implicant
 - prime implicant is essential if it alone covers an element of the ON-set
 - will participate in ALL possible covers of the ON-set
 - DC-set used to form prime implicants but not to make implicant essential
- ❑ Objective:
 - grow implicant into prime implicants (minimize literals per term)
 - cover the ON-set with as few prime implicants as possible (minimize number of product terms)

Examples to illustrate terms



6 prime implicants:

$A'B'D, BC', AC, A'C'D, AB, B'CD$

essential

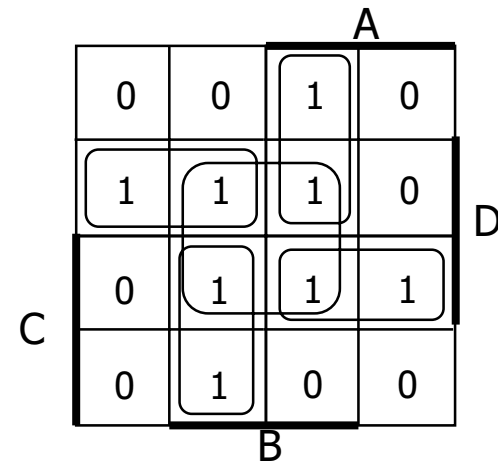
minimum cover: $AC + BC' + A'B'D$

5 prime implicants:

$BD, ABC', ACD, A'BC, A'C'D$

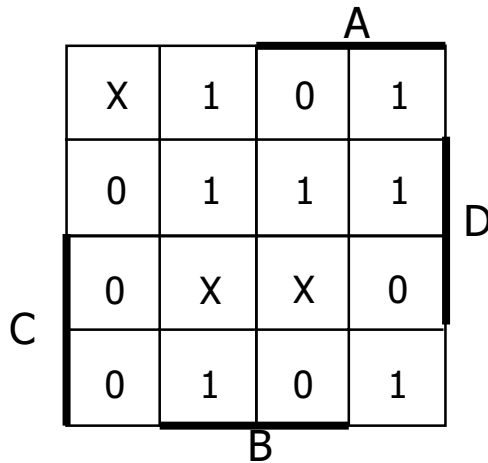
essential

minimum cover: 4 essential implicants

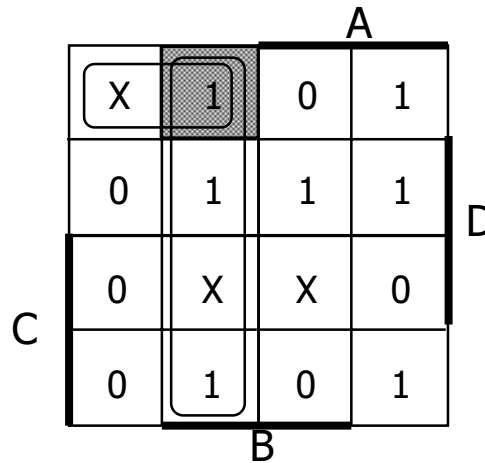


Algorithm for two-level simplification (example)

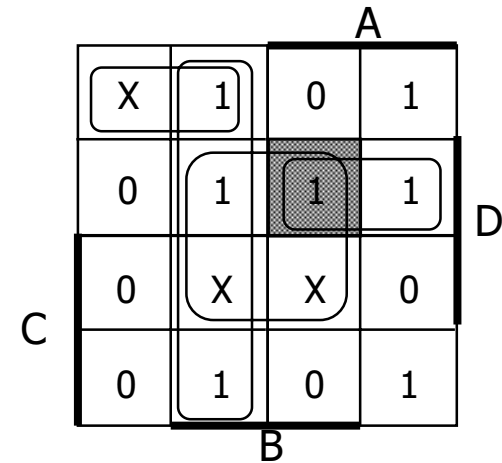
1. Find all Primes



2. Find Essentials

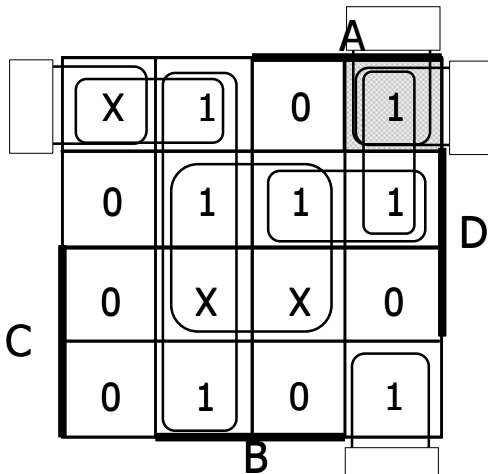


3. Find a Cover

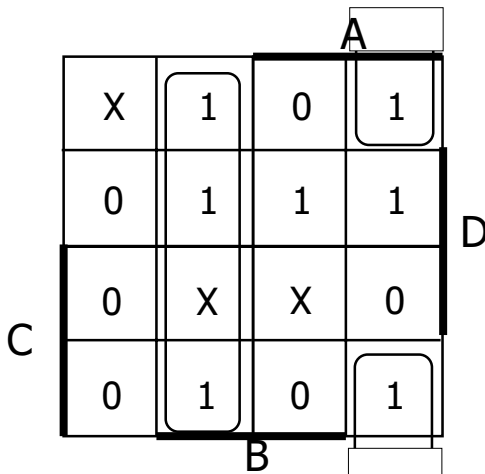


2 primes around $A'BC'D'$

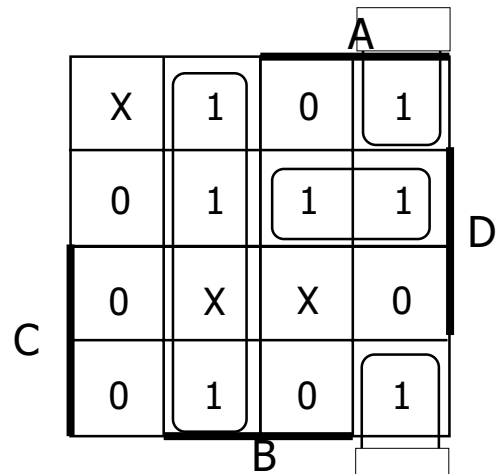
2 primes around $ABC'D$



3 primes around $AB'C'D'$



2 essential primes

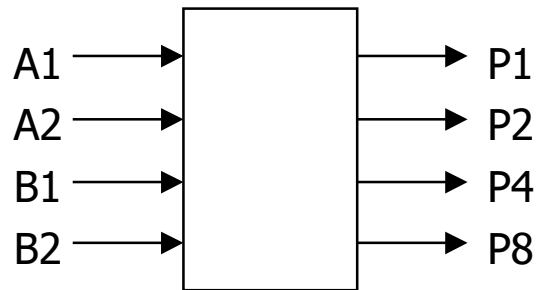


minimum cover (3 primes)

Combinational logic summary

- ❑ Logic functions, truth tables, and switches
 - NOT, AND, OR, NAND, NOR, XOR, . . ., minimal set
- ❑ Axioms and theorems of Boolean algebra
 - proofs and simplification
- ❑ Gate logic
 - networks of Boolean functions and their time behavior
- ❑ Canonical forms
 - two-level and incompletely specified functions
- ❑ Simplification
 - two-level simplification
- ❑ Later
 - multi-level logic
 - design case studies
 - time behavior

Design example: 2x2-bit multiplier



block diagram
and
truth table

A2	A1	B2	B1	P8	P4	P2	P1
0	0	0	0	0	0	0	0
		0	1	0	0	0	0
		1	0	0	0	0	0
		1	1	0	0	0	0
0	1	0	0	0	0	0	0
		0	1	0	0	0	1
		1	0	0	0	1	0
		1	1	0	0	1	1
1	0	0	0	0	0	0	0
		0	1	0	0	1	0
		1	0	0	1	0	0
		1	1	0	1	1	0
1	1	0	0	0	0	0	0
		0	1	0	0	1	1
		1	0	0	1	1	0
		1	1	1	0	0	1

4-variable K-map
for each of the 4
output functions

Design example: 2x2-bit multiplier (cont'd)

